

A Hybrid Technique for Upward Stabilization and Control of Two Wheeled Self-Balancing Segway

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ABSTRACT

Two wheeled Self-Balancing Segway, which works on the principle of inverted pendulum, is naturally unstable, nonlinear, and under actuated system. Self-Balancing means the capability of the Segway to balance itself on two wheels without falling. Therefore, the system has to be controlled to reach stability in this unstable state. The two wheeled Segway is considered important due to its applications in daily life. In this paper, a hybrid control system is proposed for upward stabilization and control of two wheeled Segway. The control design approaches proposed in the literature for Segway result in a large control effort which requires high torque causing the saturation of actuator field and ultimately failure of the controller. Controller designed using the proposed approach is able to reduce the control effort by 64% compared to the ones available in literature. Moreover, controller designed through the proposed approach is able to improve disturbance rejection for both pitch and yaw angles of the Segway. Simulation results illustrate the effectiveness of the proposed approach over the ones available in the literature.

Keywords: Segway, Hybrid Control Design, Control Effort, Saturation, Disturbance Rejection

1. INTRODUCTION

Segway is a self-balancing vehicle and it balances the person riding on it [1]. Nowadays, Segway is commonly used as a mean of transport for law enforcement, city sightseeing tours and professionals working in factories [2].

Fig. 1 shows a schematic diagram of a Two wheeled Segway with the major components and directions labeled [3].

Mathematical model of a two wheeled Segway is based on that of inverted pendulum [4-6]. Two wheeled Segway is a nonlinear, unstable system having uncertain parameters and thus poses a great deal of challenges while designing the control system. A comprehensive design of Two Wheeled Mobile Robot

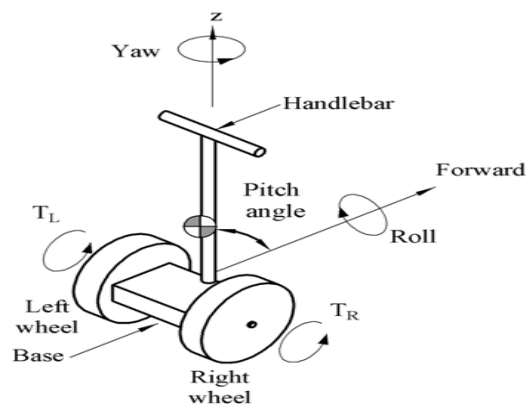


Fig. 1: Schematic Diagram of Two Wheeled Segway

(TWMR) are addressed in [7-8]. Issues related to selection of sensors and actuators, control scheme, signal processing units and modelling are addressed in the afore-mentioned work. A simple control scheme for

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the Segway, based on pole placement method, has also been proposed there. Lin *et al.* [9] applied a simple PID control and state-feedback control using pole placement method for the Segway human transporter. Takei and Imamura [10] presented Linear Quadratic Regulator (LQR) approach for pitch angle control of self-balancing Segway. The approaches discussed above are based on nominal model and can only ensure the desired performance when the parameters of Segway remain constant.

Aside from the Linear Time Invariant (LTI), nonlinear methods are also proposed for the Segway in the literature. Yau *et al.* [11] and Wu *et al.* [12] proposed model for the two wheeled self-balancing robot and designed a sliding mode controller. The sliding mode combined with robust control is proposed for stabilization and disturbance rejection of two wheeled self-balancing robot. However, this also resulted in chattering in control effort due to the switching between sliding surfaces. Kokotovic [13] and Dumitrascu *et al.* [14] proposed the back stepping technique for the stabilization of two wheeled, they combined back stepping with sliding mode control to suppress the chattering in control effort.

Kausar *et al.* [15] studied the movement of two wheeled mobile Segway on inclined terrain to avoid the tip-over issue. The research work explained the adjustment for center of mass of the body to deal with the problem in balancing of Segway on an inclined surface. The LQR method was applied to design the controller. The controller was then tested for inclined and flat surfaces. The results revealed that for varying equilibrium point and stability region, the stability region of Segway motion on inclined terrain was reduced.

Park *et al.* [16] proposed a technique based upon adaptive neural sliding mode control for handling uncertainties in non-holonomic model of two wheeled self-balancing Segway robot. Uncertainties in model and external disturbances were approximated by using Self-Recurrent Wavelet Neural Networks (SRWNNs). The simulation results revealed the robustness and better performance of the proposed designed control system.

Tsai *et al.* [17] presented an adaptive control technique

based on Radial Basis Function Neural Networks (RBDNNs) for two wheeled Segway. The pitch angle and yaw controls were achieved by adaptive controllers using RBFNNs. However, the NNs require a lot of time for training and in turn a large memory bank is required and also they may stuck up in local minima.

In the work of Pham and Lee [18] and Shilpa *et al.* [19], sliding mode control was applied for stability of two wheeled Segway. The advantage of sliding mode control technique was its insensitivity to the parameter uncertainties and modeling errors of the system.

Rashdi *et al.* [20] proposed a nonlinear control technique based on SMC for pitch, yaw, and altitude control of a quad copter. The robustness of the SMC design provided a better control design having good tracking performance as well as accuracy [20]. Similar approaches are also applied to the Segway, as mentioned above.

Son and Anh [21] proposed the technique of back stepping to control the tilt angle of the Segway. The main drawback of this technique was its complexity and results revealed that the control input effort given to the plant by the controller was very high. This problem is serious as the input depends upon the specifications of motors. The torques of the motors are limited and may cause saturation resulting in failure of the designed controller.

It is clear from the above discussion that the LTI control design approaches render quite large control effort while the nonlinear approaches result in chattering in the control effort. The chattering can be removed by combining the conventional nonlinear control design approaches, *i.e.* System Management Controller (SMS) or Backstepping, with the neural network but the resulting algorithm becomes computationally expensive. To avoid all these issues, a hybrid control approach using conventional LTI approaches has been proposed in this work. The major contributions of this work include:

- Application and validation of the proposed hybrid approach on Segway for stability and disturbance rejection
- Stabilization and disturbance rejection at a

reduced control effort with no chattering

- Uncertainties due to the rider's mass have been compensated

Rest of the article is arranged as: Section 2 described the detailed mathematical modeling of the Segway. Section 3 discusses the proposed hybrid control design approach for the Segway. A detailed discussion about the simulation results have been provided in Section 4 and finally conclusions are drawn in Section 5.

2. MATHEMATICAL MODELING OF THE SEGWAY

This section describes the mathematical model of the two wheeled Segway. Basic laws of physics and Newton's method are applied to determine the mathematical equations of the Two-wheeled Segway. Coordinate system of the Segway is shown in Fig. 2. Segway model is derived by applying Newton's second law of motion on the system. Subscripts L stands for left and R for right sides of the Segway.

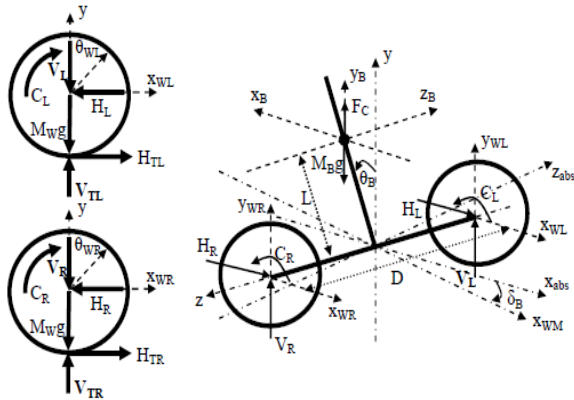


Fig. 2: Two Wheeled Segway Coordinate System

Equations (1-6) are for left wheel of the Segway (same for the right wheel).

$$M_W \ddot{x}_{WL} = H_{TL} - H_L \quad (1)$$

$$M_W \ddot{y}_{WL} = H_{TL} - H_L \quad (2)$$

$$J_{WL} \ddot{\theta}_{WL} = C_L - H_{TL}R \quad (3)$$

$$x_{WL} = \theta_{WL}R \quad (4)$$

$$J_{WL} = \frac{1}{2} m_{WL} R^2 \quad (5)$$

$$\delta = \frac{x_{WL} - x_{WR}}{D} \quad (6)$$

Equations (7-17) provide mathematical representation of the Segway body.

$$M_B \ddot{x}_{WB} = H_L + H_R \quad (7)$$

$$M_B \ddot{y} = V_L + V_R - M_B g + \frac{C_L + C_R}{L} \sin \theta_B \quad (8)$$

$$J_B \ddot{\theta}_B = (V_L + V_R)L \sin \theta_B - (H_L + H_R)L \cos \theta_B - (C_L + C_R) \quad (9)$$

$$x_B = L \sin \theta_B + \frac{x_{WL} + x_{WR}}{2} \quad (10)$$

$$y_B = -L(1 - \cos \theta_B) \quad (11)$$

Where distance between the center of the gravity of Segway and body pitch inertia can be calculated as given in [22].

$$L = \left(M_S \left(\frac{H_B}{2} \right) + L_{rider} \times 0.55 \times M_{rider} \right) / M_B \quad (12)$$

$$J_{rider} = \frac{1}{3} M_{rider} L_{rider}^2 \quad (13)$$

$$J_B = J_{rider} + 1.49 \quad (14)$$

$$\theta = \theta_B = \theta_w = \theta_{wL} = \theta_{wR} \quad (15)$$

$$x_{wM} = \frac{x_{wL} + x_{wR}}{2} \quad (16)$$

$$J_\delta \ddot{\delta} = \frac{D}{2} (H_L + H_R) \quad (17)$$

where HTR, HTL, HR, HL, VTR, VTL, VR, and VL denote the reaction forces on the different free bodies.

Substituting equations (7, 8, 15) into equation (9) gives equation (18)

$$J_B \ddot{\theta} = M_B (\ddot{y} \sin \theta - \dot{x}_B \cos \theta) + M_B g L \sin \theta - (C_L + C_R)(1 + \sin^2 \theta) \quad (18)$$

Equation (19) is derived from equations (10-11, 16),

$$\ddot{y}_B \sin \theta - \dot{x}_B \cos \theta = -L \ddot{\theta} - \dot{x}_{wM} \cos \theta \quad (19)$$

Equation (20) is derived from substitution of equations (14, 17) into equation (18).

$$\frac{4}{3} M_B L^2 \ddot{\theta} + M_B L \cos \theta \ddot{x}_{wM} = M_B g L \sin \theta - (a + \sin^2 \theta) C_\theta \quad (20)$$

Manipulating equation (1) gives equation (21)

$$M_W (\ddot{x}_{wL} + \ddot{x}_{wR}) = -(H_L + H_R) + (H_{TL} + H_{TR}) \quad (21)$$

Substituting equations (3, 7) into equation (21) and simplifying.

$$M_W(x_{WL}'' + x_{WR}'') = -M_B \ddot{x} + \frac{C_L + C_R - (J_{WL}\ddot{\theta}_{WL} + J_{WR}\ddot{\theta}_{WR})}{R} \quad (22)$$

Equation (23) is derived from Equations (10, 14)

$$x_B'' = \ddot{\theta}L\cos\theta + \dot{\theta}L\cos\theta + x_{WM}'' \quad (23)$$

Substituting equations (23, 5) into equation (22)

$$(M_B L \cos\theta + M_W R)\ddot{\theta} + (2M_W + M_B)x_{WM}'' = \dot{\theta}^2 M_B L \sin\theta + \frac{C_\theta}{R} \quad (24)$$

From Equations (20,24), we have equations (25-26)

$$A\ddot{\theta} = B_1\dot{\theta}^2 + C_1 C_\theta \quad (25)$$

$$A x_{WM}'' = B_2 \dot{\theta}^2 - C_2 C_\theta \quad (26)$$

Equation (27) is derived from equations (1,3-4)

$$H_L = \frac{C_L}{R} - x_{WL}'' [M_W + \frac{J_{WL}}{R^2}] \quad (27)$$

From equation (6), we get:

$$\ddot{\delta} = \frac{x_{WL}'' - x_{WR}''}{D} \quad (28)$$

From equations (27-28), we obtain:

$$H_L - H_R = \frac{C_L - C_R}{R} - D\ddot{\delta} [M_W + \frac{J_W}{R^2}] \quad (29)$$

Substituting equation (29) into equation (17), we have

$$\left[J_\delta + \frac{1}{2} D^2 \left[M_W + \frac{J_W}{R^2} \right] \right] \ddot{\delta} = \frac{1}{2} D \frac{C_L - C_R}{2} \quad (30)$$

$$J_W = \frac{1}{2} M_W R^2 \text{ and } J_\delta = \frac{1}{3} M_B \left[\frac{D}{2} \right]^2 = \frac{1}{12} M_B D^2 \quad (31)$$

Substituting equation (31) into equation (30), we have equation (32)

$$\ddot{\delta} = C_3 C_\delta \quad (32)$$

State space representation of the Segway as a non-linear system can be described by equations (25, 26, 32), where total torque required to change pitch angle is $C_\theta = C_L + C_R$ and total torque required to change yaw angle is $C_\delta = C_L - C_R$.

$$A = 2M_g + M_B - \frac{0.75(M_W R + M_B L \cos\theta) \cos\theta}{L} \quad (33)$$

$$B_1 = \frac{0.75g(2M_W + M_B) \sin\theta}{L} - \frac{0.75M_B L \sin\theta \cos\theta}{L} \dot{\theta}^2 \quad (34)$$

$$C_1 = - \left(\frac{0.75(1 + \sin^2\theta)(2M_W + M_B)}{M_B L^2} + \frac{0.75 \cos\theta}{RL} \right) \quad (35)$$

$$B_2 = \frac{-0.75g(M_W R + M_B L \cos\theta) \sin\theta}{L} + M_B L \sin\theta \dot{\theta}^2 \quad (36)$$

$$C_2 = \frac{-0.75(M_W R + M_B L \cos\theta)(1 + \sin^2\theta)}{M_B L^2} + \frac{1}{R} \quad (37)$$

$$C_3 = \frac{6}{(9M_W + M_B)RD} \quad (38)$$

The Nonlinear model of Segway is linearized at $x=0$. The Segway is considered in upright at the point $x=0$. The Linearized equations are as follows:

$$\dot{x}_1 = x_4 \quad (39)$$

$$\dot{x}_2 = Qx_3 - 2SC \quad (40)$$

where x_1 = Position of the Segway, x_2 = Velocity of the Segway, x_3 = Pitch Angle of the Segway

x_4 = Angular velocity of the Segway

$$P = \frac{0.75(2M_W + M_B)}{2M_W L + M_B L - 0.75(M_W R + M_B L)} \quad (41)$$

$$Q = \frac{-0.75g(M_W R + M_B L)}{(2M_W L + M_B L) - 0.75(M_W R + M_B L)} \quad (42)$$

$$R = \frac{0.75R(2M_W + M_B) - (0.75M_B L)}{M_B L R (2M_W L + M_B L - 0.75(M_W R + M_B L))} \quad (43)$$

$$S = \frac{0.75(M_W R + M_B L)R + M_B L^2}{M_B L R (2M_W L + M_B L - 0.75(M_W R + M_B L))} \quad (44)$$

State Space equations of the Linearized model of the Segway system are

$$\dot{x} = Ax + Bu \quad (45)$$

$$y = Cx + Du \quad (46)$$

where

$$A = \begin{bmatrix} 0 & 1 & 0 & 0 \\ 0 & 0 & Q & 0 \\ 0 & 0 & 0 & 1 \\ 0 & 0 & P & 0 \end{bmatrix}, \quad (47)$$

$$B = \begin{bmatrix} 0 \\ -2SR \\ 0 \\ 2R \end{bmatrix} \quad (48)$$

$$C = \begin{bmatrix} 1 & 0 & 0 & 0 \\ 0 & 0 & 1 & 0 \end{bmatrix} \quad (49)$$

$$D = \begin{bmatrix} 0 \\ 0 \end{bmatrix} \quad (50)$$

$$x = \begin{bmatrix} x_1 \\ x_2 \\ x_3 \\ x_4 \end{bmatrix} \quad (51)$$

Symbol	Value [Unit]	Parameters
θ	[rad]	Pitch Angle of Segway
δ	[rad]	Yaw angle of Segway
M_w	7[kg]	Wheel mass
M_{rider}	80[kg]	Rider mass
M_s	26 [kg]	Segway mass
M_B	[kg]	Total body mass ($M_B=M_s+ M_{rider}$)
R	0.2[m]	Wheel radius
L	[m]	Distance between the center of the gravity of Segway and z axis
D	0.6[m]	Distance between the contact points of the two wheels
g	9.8[m/s ²]	Constant of Gravity
H_B	0.03[m]	Body height
L_{rider}	1.8[m]	Rider height
J_{rider}	[kgm ²]	Rider inertia
J_B	[kgm ²]	Body pitch inertia
C_L, C_R	[N.m]	Torques at the inputs of left and right wheels
H_{TL}, H_{TR}	[N]	Friction between the wheels and ground surface
H_L, H_R	[N]	Impact of Reaction forces on the left and right wheels.
J_{TL}, J_{TR}	[kgm ²]	Moment of Inertia of the rotating masses with reference to the z axis
θ_{WL}, θ_{WR}	[rad]	Angle of the left and right wheels

3. CONTROLLER DESIGN

The main features of the proposed hybrid control strategy are described as follows:

- Design of mixed sensitivity-based H_∞ robust controller to keep the Segway in equilibrium position having tilt angle $\theta = 0^\circ$, *i.e.* to ensure the upward stabilization of human rider.
- A right and left-turning control system designed to control the Segway in turning right and left. In this paper, an LQR control system is used to design a right- and left-turning of the Segway.

3.1 H_∞ Control System Design

H_∞ controller design methodology has a number of advantages which include stability, disturbance rejection, robustness and it also caters modelling errors. Design specifications such as uncertainty in model [23], reference tracking at lower frequencies and disturbance rejection at higher frequencies are generally addressed in H_∞ controller design techniques. This technique provides a close loop response of the system based on the bandwidth which is specified by appropriate performance weights selection. H_∞ controller design technique incorporates the weights which depend on frequency and are responsible for shaping the closed loop system's response.

In this work, the rider's mass has been taken as a modeling uncertainty because the exact mass of the rider is unknown at the time, *i.e.* different riders possess different mass.

3.2 Generalized Problem Formulation

Feedback controller design may have many control configurations but it is useful to have a generalized configuration so that every problem can be formulated according to the requirements [24]. Such a generalized control configuration is shown in Fig. 3.

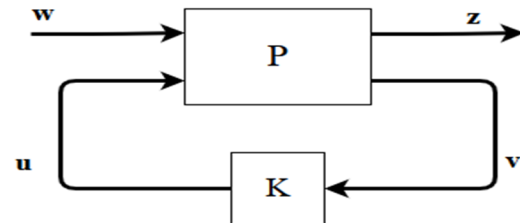


Fig. 3: Generalized Control Configuration

Generalized control configuration of the system is

$$\begin{bmatrix} z \\ v \end{bmatrix} = P(s) \begin{bmatrix} w \\ u \end{bmatrix} = \begin{bmatrix} P_{11}(s) & P_{12}(s) \\ P_{21}(s) & P_{22}(s) \end{bmatrix} \quad (52)$$

$$u = K(s)v \quad (53)$$

where u = variables to control., v = variables to be measured, w = outer signals disturbances and references, z = signals to be minimized to achieve the control objectives.

The linear fractional transformation gives the closed loop transfer function between w and z .

$$z = F_1(P, K)w \quad (54)$$

where

$$F_1(P, K) = P_{11} + P_{12}K(I - P_{22}K)^{-1}P_{21} \quad (55)$$

The H_∞ controller minimizes the H_∞ norm of the lower fractional transform $F_1(P, K)$.

3.3 Generalized H_∞ Loop Shaping Design

There are multiple methods for designing of H_∞ controller but loop shaping design is most popular because of its simple design procedure and ability to deal with a broad class of uncertainties [23]. Postlethwaite and Skogestad [24] proposed a design of loop shaping for H_∞ controller. This technique is a combination of loop shaping and robust stabilization.

There are two steps involved in loop shaping design procedure.

- (1) Augmentation of plant with pre and post compensator.
- (2) Stabilization of shaped plant using coprime uncertainty optimization

Loop shaping design is preferred over other techniques of H_∞ controller design, as robust stabilization problem does not require any γ iterations to reach at a solution [24]. Moreover, the weight selection or problem dependent uncertainty is not required for the design. As there are no weights on the perturbations so it is reasonable for the nominal plant to use normalized coprime factorization. This description of uncertainty is general and is advantageous when we do not have specific information about the uncertainties in the plant as it represents a broad class of uncertainties and allows the movement of both zeroes and poles from left half plane to right half plane.

3.4 Mixed Sensitivity H_∞ Control

In mixed sensitivity, we used sensitivity function which mathematically written as $S = (I - GK)^{-1}$ is shaped for closed loop transfer functions for example KS or the complementary transfer function $T = I - S$ and such type of problems are called mix sensitivity optimization shaping problems as shown in Fig. 4.

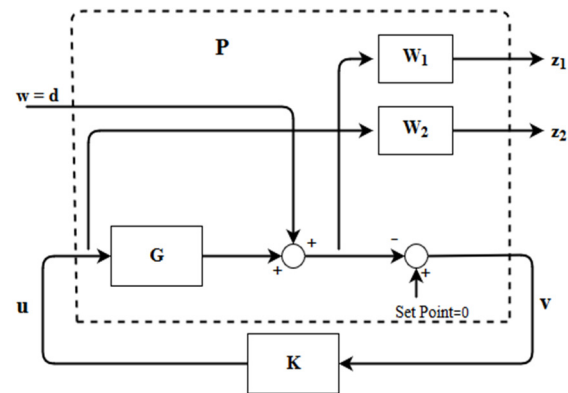


Fig. 4. H_∞ Mixed Sensitivity in Standard Form (Disturbance Rejection)

$w_1(s)$ = Scalar low pass filter
 $w_2(s)$ = Scalar high pass filter

$$\left\| \begin{matrix} w_1 S \\ -w_2 KS \end{matrix} \right\|_\infty \quad (56)$$

In General, mixed sensitivity problem can be expressed as:

$$\begin{aligned} \text{Error signal: } z &= [z_1^T \ z_2^T]^T \\ z_1 &= W_1 y \\ z_2 &= -W_u u \end{aligned}$$

According to figure configuration of (3).

$$\begin{aligned} z_1 &= W_1 S w \\ z &= W_2 K S w \end{aligned}$$

So elements of generalized power plant P are:

$$P_{11} = \begin{bmatrix} W_1 \\ 0 \end{bmatrix} \quad (57)$$

$$P_{12} = \begin{bmatrix} W_1 G \\ -W_2 \end{bmatrix} \quad (58)$$

$$P_{21} = -I \quad (59)$$

$$P_{22} = -G \quad (60)$$

H_∞ controller minimizes the H_∞ norm of the fractional transform $F_l(P, K)$.

$$F_l(P, K) = \begin{bmatrix} W_1 S \\ -W_2 K S \end{bmatrix} \quad (61)$$

3.5 LQR Controller to Control the Yaw Angle of Segway

Linear Quadratic Regulator (LQG) technique is

is related to the theory of optimal control in which a cost function is to be minimized to solve LQ problems. This technique is used with some other techniques like Routine Kalman Filtering to get best control response for the naturally unstable systems [25, 26].

To control the Yaw angle of the Two Wheeled Segway, LQR-control can be applied. The general configuration of LQR is shown in Fig. 5.

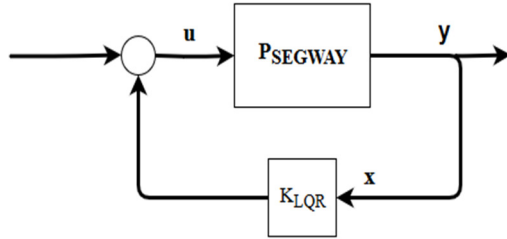


Fig. 5: Block Diagram of LQR Controller for the System (Segway)

The cost function of the performance index is given by

$$J = \int_0^{\infty} (X^T Q X + u^T R u) dt \quad (62)$$

For LQR Controller design, the control effort, u_{input} is given as:

$$u_{input} = -Kx \quad (63)$$

Gain matrix of K is defined as

$$K = R^{-1} B^T P \quad (64)$$

And P is solved from the Ricatti Equation

$$Q + A^T P + P A - P B B^T P = 0 \quad (65)$$

Since the state space system is controllable and observable, P has the only unique solution when the state space system is observable as well as controllable, so that the closed loop system poles are strictly in the right half plane. However, it's necessary to make it clear that for a particular change in input, for example an impulse on the Segway, that the control effort u_{input} determine by the controller is lie within some certain bounds.

Using Equation (36), the transfer function of Yaw angle of the Segway is described as;

$$G_{yaw}(s) = \frac{\delta(s)}{C_{\delta}(s)} = \frac{1}{s^2} C_3 \quad (66)$$

4. PROPOSED CONTROL DESIGN FOR THE SEGWAY

The proposed hybrid control system for two wheeled Segway has been simulated using MATLAB Simulink. Block diagram of the Hybrid controller design is shown in Fig. 6.

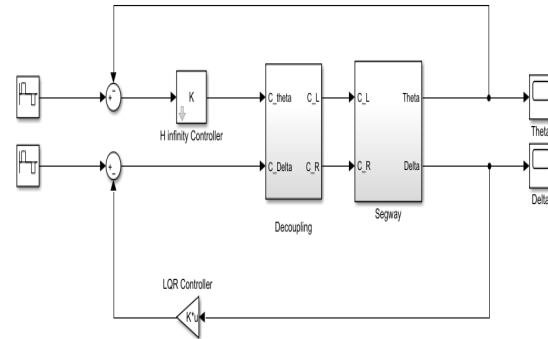


Fig. 6: Block Diagram of the Hybrid Segway Controller

The Controller K (H_{∞}) is achieved as:

$$K = \frac{-2.956e08 s^3 - 8.01e08 s^2 - 2.942e09 s - 7.973e09}{s^4 + 1.616e06 s^3 + 3.172e07 s^2 + 5.033e07 s + 2.086e07} \quad (67)$$

Weights for H_{∞} Mix Sensitivity Loop Shaping are:

$$W_p = \frac{s^2 + 10.44 s + 27.25}{0.007569 s^2 + 0.01285 s + 0.00545} ; W_u = 1 \quad (68)$$

Disturbance Transfer Function is:

$$G_d = \frac{10}{50 s + 500} \quad (69)$$

5. RESULTS AND DISCUSSIONS

First of all, we designed the S function block to simulate the non-linear response of two wheeled Segway. The results are shown in Figs. 7-12.

5.1 Open loop analysis

In Figs. 7-8, the analysis shows that the open loop system is unstable. So, it is required to design a control system to stabilize the response for both pitch and yaw angles.

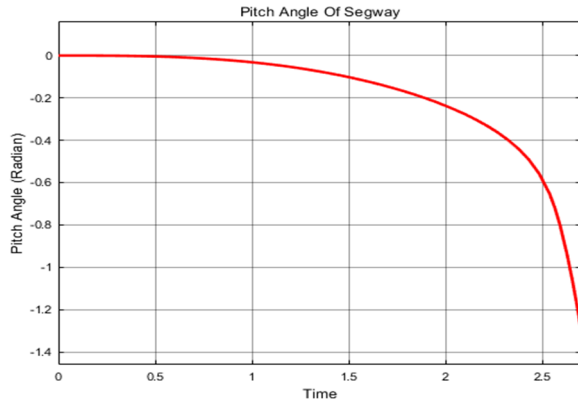


Fig. 7: Step Response of the Segway Pitch Angle

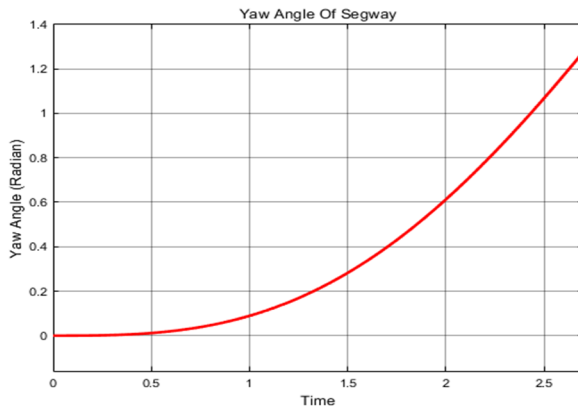


Fig. 8: Step Response of the Segway Yaw Angle

5.2 Results of Hybrid Control System

In Fig. 9, the simulation results illustrate that in response of the square wave disturbance, which is to the tilt angle of the Segway, the Segway angle disturbs to only 0.1 rad/s and comes back to its equilibrium value within less than 0.5 seconds. The overshoot is very small.

In Fig. 10, the control effort of the controller in Nm reaches only 18 Nm which is very less as compared to the back- stepping technique. This ensures the effectiveness of H_{∞} controller.

The simulation results in Fig. 11 illustrate that the settling time is almost 0.2 seconds for LQR Control design. So, the overall Hybrid controller design shows the reasonable results in terms of both disturbance rejection and control effort. All the simulations are performed for 0.1 radian change in Segway Pitch and Yaw angles. The Yaw angle of the Segway is also controlled effectively using LQR controller.

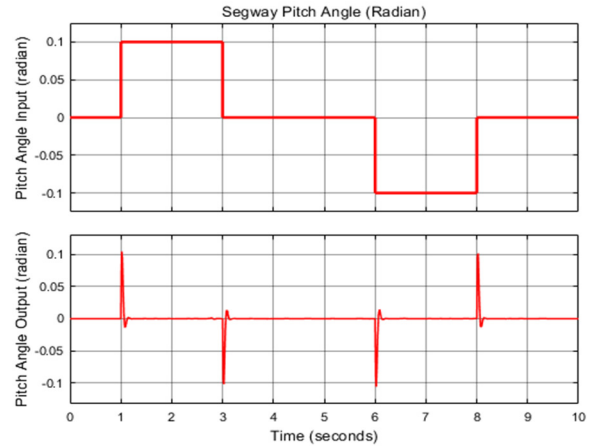


Fig. 9: Segway Pitch Angle Response for 0.1 Radian Change

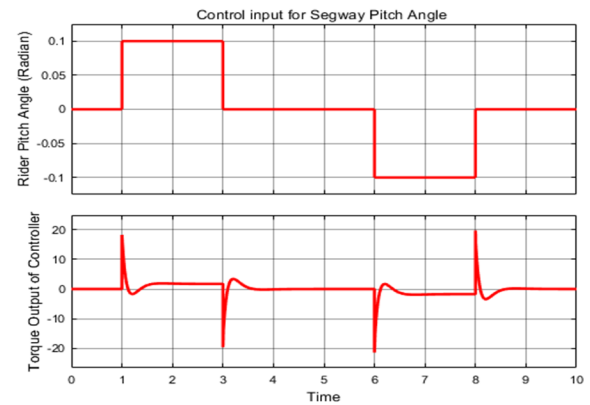


Fig. 10: Torque Output of Pitch Angle Controller for 0.1 Radian Change

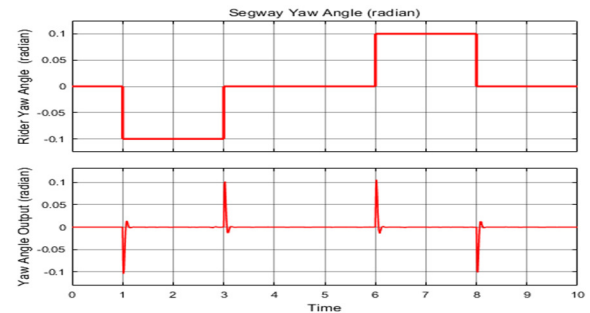


Fig. 11: Segway Yaw Angle Response for 0.1 Radian Change

5.3 Relationship between Torque Difference and Yaw Angle

A right and left-turning control system is designed to control the Segway in turning right and left. A change in yaw angle is required to turn Segway left and right. According to Fig. 12, when torque applied on left

wheel (C_L) is greater than right wheel (C_R) then variation in delta is positive so Segway will move to right side. Similarly, when torque applied on right wheel (C_R) is greater than left wheel (C_L) then variation in delta is negative so Segway will move to left side.

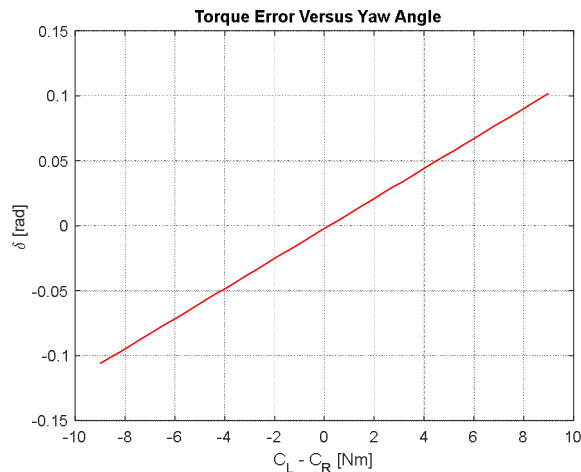


Fig. 12: Relationship between Torques Vs Yaw Angle (Delta)

The comparison between the proposed hybrid approach and Backstepping control strategy is given in Table 2. The Table 2 shows that the control effort achieved by applying the proposed technique is much lesser compared to that of using Backstepping in [21].

Table 2: Comparisons between Backstepping and Proposed Technique			
Technique	Peak Value (rad)	Settling Time (Seconds)	Control Input (Nm)
Backstepping	0.1	2.5	50
Hybrid Technique	0.1	0.5	18

6. CONCLUSION

A hybrid technique for the stabilization and control on one of the applications of inverted pendulum i.e. two wheeled Segway is successfully applied. Many techniques including LQR, sliding mode control, back stepping etc. are available for controlling two wheeled Segway, where every technique has its own pros and cons. Out of all these approaches, back-stepping is one of the main approaches used for upward stabilization of two wheeled Segway. The main drawback of this technique is its complexity and increased control

effort.

To resolve this problem, a robust control *i.e.* H_∞ controller using mix sensitivity loop shaping for controlling the pitch angle of the Segway and the LQR controller to control the Yaw Angle of the Segway is combined to use to benefits of both. The simulation results of our proposed technique reveal that the disturbance rejection has been improved for a much lesser control effort compared to back stepping technique [21] and thus avoids the saturation of motors. Reducing control effort is helpful to avoid the saturation of motors which causes the failure of control system. It is a significant achievement to reduce the control effort from 50 to 18 Nm for the Segway, which is almost 64%. In future, by utilizing this hybrid technique, one can apply this approach on a real time system ensuring the reliability and increased the life span of actuators used in two wheeled Segway.

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